

Identification of Broadband Source-Array Responses from Sensor Second Order Statistics

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Presentation Overview



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- 6. Conclusions



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Space-Time Covariance

- We have M sensor signals organised in $\mathbf{x}[n] \in \mathbb{C}^M$;
- ► to take the broadband nature of signals into account, we must consider lags *τ*;
- space-time covariance matrix $\mathbf{R}[\tau] = \mathcal{E}\{\mathbf{x}[n]\mathbf{x}^{\mathrm{H}}[n-\tau]\}$;





Cross-Spectral Density Matrix

 CSD matrix forms a *z*-transform pair with the space-time covariance matrix,

$$\boldsymbol{R}(z) = \sum_{\tau} \mathbf{R}[\tau] z^{-\tau}$$
 or $\boldsymbol{R}(z) \bullet \mathbf{R}[\tau];$

• symmetry of $\mathbf{R}[\tau] \longrightarrow \mathbf{R}(z)$ is parahermitian:

$$\boldsymbol{R}(z) = \boldsymbol{R}^{\mathrm{P}}(z) = \boldsymbol{R}^{\mathrm{H}}(1/z^{*});$$

(Hermitian transposition and time reversal)

• link to a narrowband covariance at normalised angular freq. Ω_k ,

$$\mathbf{R}(\mathrm{e}^{\mathrm{j}\Omega_k}) = \mathbf{R}(z)\big|_{z=\mathrm{e}^{\mathrm{j}\Omega_k}}$$

 many optimal (narowband!) methods are based on decompositions such as the EVD: R(e^{jΩk}) = QΛQ^H.



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McWhirter Decomposition



 $\boldsymbol{R}(z) \approx \boldsymbol{Q}(z) \boldsymbol{\Lambda}(z) \boldsymbol{Q}^{\mathrm{P}}(z)$

▶ paraunitary (i.e. lossless) matrix Q(z), s.t. $Q(z)Q^{P}(z) = I$;

• diagonal and spectrally majorised $\mathbf{\Lambda}(z)$:





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Parahermitian Matrix EVD (PEVD)



- Franz Rellich (1937): for R(e^{jΩ}) analytic, there exist analytic eigenvectors Γ(e^{jΩ}) and analytic eigenvalues U(e^{jΩ});
- can be generalised to

$$\boldsymbol{R}(z) = \boldsymbol{U}(z)\boldsymbol{\Gamma}(z)\boldsymbol{U}^{\mathrm{P}}(z) ;$$



▶ eigenvalues are unique, eigenvectors can be modified by arbitrary allpass filters H(z) (s.t. $H(z)H^{\rm P}(z) = 1$),

$$\boldsymbol{R}(z)\boldsymbol{u}(z)H(z) = \gamma(z)\boldsymbol{u}(z)H(z)$$
.

Source-Sensor Transfer Functions

▶ We take *M*-array measurements of a single source:







Transfer Functions and PEVD



- ▶ 2nd order stats: $R_i(z) = S(z)a_i(z)a_i^{\mathrm{P}}(z) = \gamma_{i,m}(z)u_i(z)u_i^{\mathrm{P}};$
- ▶ difference: $u_i(z)$ is normal, $u_i^{\rm P}(z)u_i(z) = 1$, while $a_i(z)$ is not:

$$\boldsymbol{a}_{i}^{\mathrm{P}}(z)\boldsymbol{a}_{i}(z) = A_{i,(-)}(z)A_{i,(+)}(z) = A_{i,(+)}^{\mathrm{P}}(z)A_{i,(+)}(z)$$

with minimum-phase $A_{(+)}(z)$;

therefore:

$$\begin{split} H_i(z) \boldsymbol{u}_i(z) &= \frac{\boldsymbol{a}_i(z)}{A_{i,(+)}(z)} \\ \gamma_i(z) &= A_{i,(+)}(z) S(z) A_{i,(+)}^{\mathrm{P}}(z) \;, \end{split}$$

Form a single measurement $R_i(z)$, we cannot say anything about $a_i(z)$ or S(z).



Multiple Measurements



• If we have several measurements $i = 1 \dots I$:

$$H_i(z)\boldsymbol{u}_i(z) = \frac{\boldsymbol{a}_i(z)}{A_{i,(+)}(z)}$$

$$\gamma_i(z) = A_{i,(+)}(z)S(z)A_{i,(+)}^{\rm P}(z) ,$$

• we can extract S(z) as the greatest common divisor

$$\hat{S}(z) = \mathsf{GCD}\{\gamma_1(z) \ \dots \ \gamma_I(z)\};\$$

► we can then extract the terms A_{i,(+)}(z), and hence determine the vectors a_i(z) save of an arbitrary phase response due to the allpass H_i(z):

$$\boldsymbol{a}_i(z) = A_{i,(+)}(z)H_i(z)\boldsymbol{u}_i(z) \; .$$



Alternative DFT Domain Attempt



• As an alternative, we take measurements in independent frequency bins $\Omega_k = \frac{2\pi k}{K}$:

$$\begin{split} \mathbf{R}_{i,k} &= \boldsymbol{R}_i(e^{j\Omega_k}) = \boldsymbol{a}_i(e^{j\Omega_k}) S(e^{j\Omega_k}) \boldsymbol{a}_i^{\mathrm{H}}(e^{j\Omega_k}) \\ &= \mathbf{q}_{i,k} \lambda_{i,k} \mathbf{q}_{i,k}^{\mathrm{H}} \; . \end{split}$$

 principal eigenvectors and eigenvalues for the measurement campaigns are

$$\mathbf{q}_{i,k} = \frac{\boldsymbol{a}_i(e^{j\Omega_k})}{|\boldsymbol{a}_i(e^{j\Omega_k})|} ,$$
$$\lambda_{i,k} = S(e^{j\Omega_k})|\boldsymbol{a}_i(e^{j\Omega_k})|^2 .$$

because of the normalisation, nothing can be extracted about the source or the transfer functions.



Numerical Example

Source with power spectral density

$$S(z) = \frac{1}{2}z + \frac{5}{4} + \frac{1}{2}z^{-1}$$

• vector of transfer functions during campaign i = 1:

$$a_1(z) = \begin{bmatrix} 1 & + & \frac{1}{2}z^{-1} \\ \frac{3}{4} & - & \frac{1}{2}z^{-1} \end{bmatrix}$$

• vector of transfer functions during campaign i = 2:

$$a_2(z) = \begin{bmatrix} \frac{4}{5} & - & \frac{1}{2}z^{-1} \\ -\frac{1}{2} & + & z^{-1} \end{bmatrix};$$

▶ based on these: PEVD computations for R₁(z) and R₂(z), and GCD calculation based on eigenvalues.



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Numerical Results — Source PSD

• Eigenvalues / source PSD for both measurements $i = \{1, 2\}$.



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Numerical Result — Magnitude Responses I

• Eigenvectors / magnitude response for measurement $i = \{1\}$.



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Numerical Result — Magnitude Responses II

• Eigenvectors / magnitude response for measurement $i = \{2\}$:



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Summary and Critique



- We can extract the source PSD and the magnitude responses once we have at least two measurements;
- an independent frequency bin approach does not yield anything;
- the polynomial approach rests on an accurate parahermitian EVD, and an accurate root finding / GCD algorithm;
- root finding is numerically challenging: research since Euclid (300BC), with robust root-finding methods still on-going (Gröbner bases, algebraic geometry);
- nevertheless the approach gives a glimpse of the type of advantages that a coherent broadband approach can offer.